



iJRASET

International Journal For Research in
Applied Science and Engineering Technology



INTERNATIONAL JOURNAL FOR RESEARCH

IN APPLIED SCIENCE & ENGINEERING TECHNOLOGY

Volume: 4

Issue: II

Month of publication: February 2016

DOI:

www.ijraset.com

Call: ☎ 08813907089

E-mail ID: ijraset@gmail.com

Improving the Power Quality Based On VSC-DSTATCOM with Back Propagation Algorithm through Neural Network

R.Bharathi¹ Dr. R.Arivalahan²

¹PG Scholar, ²Associate Professor, Department of Electrical and Electronics Engineering,
Valliammai Engineering College, Chennai

Abstract—this paper presents a three phase three wire Distributed Static Compensators (DSTATCOM) is proposed for an Power Quality improvement. DSTATCOM is a shunt device which is generally used to minimize power quality problems in distribution system. Here DSTATCOM using a Back Propagation (BP) Control Algorithm for Voltage Sag/Swell, Reactive Power Compensation, and load balancing, harmonic elimination under linear and non-linear loads. A BP-based Control Algorithm extracts reference source currents and is fed to the PWM technique to generate gate signals to the IGBT switches of VSC used DSTATCOM. Performance of the DSTATCOM is evaluated under various operating conditions using simulation model in MATLAB/SIMULINK.

Index terms—Back Propagation (BP) Control Algorithm; Distributed Static Compensator; Power Quality; Load Compensation; Load Balancing.

I. INTRODUCTION

Power quality is certainly a major issue in present era. Electrical power quality problems mainly include problems such as voltage unbalancing, flicker, harmonics, voltage sag, voltage swell, power interruption [1]. This issue is more serious in electronic systems. The harmonic level, reactive power demand are parameters that specify the degree of distortion and reactive power demand at the bus of utility [2]. Power converter-based custom power device are mostly used in reducing the power quality issues such as voltage sag and voltage swell compensation, resonance due to distortion and voltage flicker reduction within limits [3]. The performance of custom power devices depends on the control algorithm used for the estimation of reference current and pulse generation scheme. Some of the classical control algorithms p-q theory and SRF theory [3][4]. DSTATCOM has more technical advantage over SVC, they have fast response time, require less space as bulky passive components are eliminated and can be interface with real power sources such as battery, fuel cell [5]. A DSTATCOM has superior performance during low voltage condition as the reactive current can be maintained constant. The performance of the DSTATCOM depends on any control algorithm that has been applied for the extraction of reference currents. Many current control algorithm has reported for VSC based controllers [7][8]. DSTATCOM uses voltage source converters to improve productive similar to a traditional SVC and the DSTATCOM can be used to restore voltage and current. Neural network trains in Adaptive learning, self-organization, real-time operation and fault tolerance through redundant information are major advantage of these algorithm. A neural network control algorithms such as the Hopfield-type neural network is also used for the estimating the magnitude and the phase angles of the fundamental component with highly distortion voltage by the assumption of know power frequency. An improvement adaptive detecting approach for the extraction of the error signal with variable learning parameters is chosen for fast response to improve tracking speed and for low value in a stable period to improve accuracy [9]. Feedforward back propagation (BP) Artificial Neural Network (ANN) consist of various layer such as the input layer, hidden layer and output layer. It is based on feedforward BP with a high ability to deal with complex non-linear problems [10]. In this BP algorithm, the training of weights has three stages. Firstly comes the feedforward for the given inputs signal training, calculation and BP of the error signals and upgrading of training the weights[11]. It has one or more layer that is continuity, differentiability, and non decreasing monotony these are the main characteristics of this algorithm. The proposed control algorithm on DSTATCOM is implemented for the compensation of non-linear loads.

II. BASIC PRINCIPLE OF DSTATCOM

A. DSTATCOM

A DSTATCOM is a controlled reactive power, which has an voltage source inverter and a DC link capacitor connected in shunt in

International Journal for Research in Applied Science & Engineering Technology (IJRASET)

order to generate and absorb the reactive power. The terminals of the VSC in AC system are connected to the point of common coupling through an inductance which will filter the leakage inductance of coupling transformer. The DC side converter is connected and the input ripple current of the converter is controlled by DC Capacitor and is the reactive power, energy storage element. This capacitor could be recharge by a battery source. If the output voltage of VSC is higher than the terminal voltage of AC system, the DSTATCOM operates in capacitor mode. DSTATCOM is used for voltage regulation at the PCC for the compensation should be in such a way that the supply current should lead the supply voltage and for power factor correction and the supply current should be in phase with supply voltage. DSTATCOM differs when compared with other reactive power generating devices such as shunt capacitor, static VAR compensators. These senses that the ability for system unbalance or harmonic absorption. DSTATCOM has some of the benefits such as increasing power transfer capability, improved voltage stability, improved power factor, improved grid voltage control, elimination flicker, voltage balancing, power factor correction. Some of the major application where DSTATCOM used is power quality means flicker mitigation and grid voltage support.

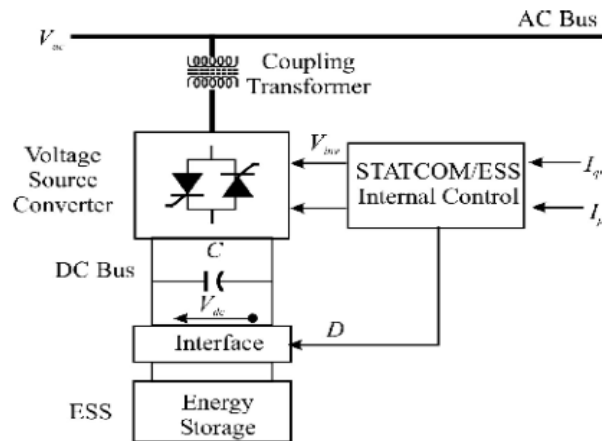


Fig 1. Basic structure of DSTATCOM

B. Operation And Configuration

DSTATCOM is a three phase and shunt power electronics based device and it is connected close to the load at the distribution systems. DSTATCOM is viewed as a variable current source obtained by a control function in order to improve the dynamic rating of capacitive range as a fixed capacitor or filter is connected in parallel with DSTATCOM. A energy storage device on DC side is for exchanging the real power with the network for a limited time during voltage sag. DSTATCOM has three modes of operation. Mode 1 operation no load mode $V_s = V_i$ then $I = 0$. Mode 2 operation is inductive mode $V_s > V_i$ then I is said to be have leading current. The magnitude of the current can be controlled by adjusting V_i here DSTATCOM will act as an capacitive reactance. Mode 3 operation is capacitive mode $V_s < V_i$ then I will said to have an lagging current, here DSTATCOM will act as inductive reactance.

C. VSC-DSTATCOM

AVSC based DSTATCOM is connected at the three phase ac mains feeding three phase linear or non linear loads with internal grid impedance. The performance of DSTATCOM depends on the accuracy of harmonic current detection. In order to reduce the ripple in compensating currents, the tuned values of interfacing inductors are connected at the output of VSC. A three phase series combination of capacitor and resistor which act as a shunt passive filter which is connected at a point of common coupling. These filter is for reducing the high frequency switching noise of VSC. The DSTATCOM currents are injected as required for the compensating currents in order to cancel the reactive power and harmonics of that load currents so that loading due to reactive power and harmonics is reduced on distribution system. The control algorithm used here is BP training algorithm, and this BP algorithm is used for the estimation of the reference source currents by calculating the weighted value of the active power and the reactive power current components. The input for this BP algorithm is, the phase PCC (V_{sa}, V_{sb}, V_{sc}), source currents (i_{sa}, i_{sb}, i_{sc}), load current (i_{La}, i_{Lb}, i_{Lc}), and dc bus voltage (V_{dc}) are required for the extraction of source current. In BP control algorithm there are two primary modes for the operation of this algorithm. They are 1. Feedforward- The information moves in the forward direction means from the input node to the hidden node and to the output node. There no cycles or loops in this feedforward. 2. BP of the

International Journal for Research in Applied Science & Engineering Technology (IJRASET)

error or supervised learning.

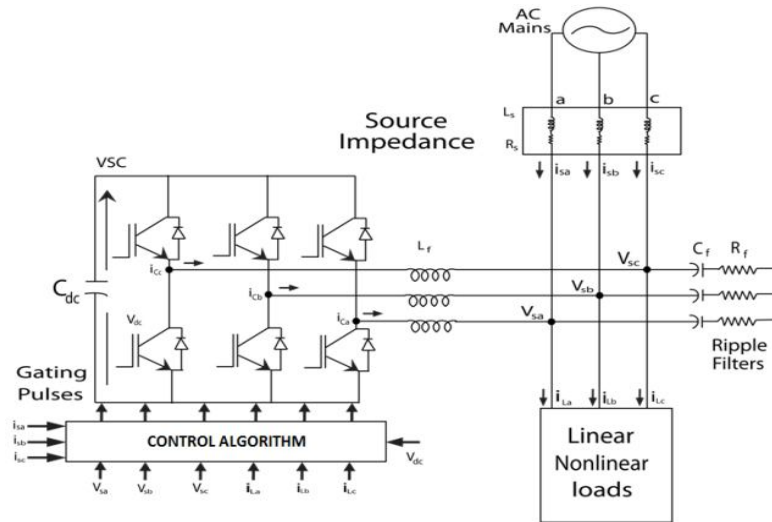


Fig 2. Schematic diagram of VSC-DSTATCOM

III. PROPOSED CONTROL ALGORITHM

This BP control algorithm for the estimation of various control parameters is as follows.

A. Weighted Function Value Of Average Active And Reactive Power Components At Fundamental

A BP training algorithm is used here for the estimation of the weighted value of load active components (W_{ap} , W_{bp} , W_{cp}) and reactive component (W_{aq} , W_{bq} , W_{cq}) from the polluted load currents using the feedforward and supervised principle. The input layer for a three phases (a,b,c) is expressed as,

$$\begin{aligned} I_{Lap} &= W_0 + i_{La}U_{ap} + i_{Lb}U_{bp} + i_{Lc}U_{cp} \\ I_{Lbp} &= W_0 + i_{Lb}U_{bp} + i_{Lc}U_{cp} + i_{La}U_{ap} \\ I_{Lcp} &= W_0 + i_{Lc}U_{cp} + i_{La}U_{ap} + i_{Lb}U_{bp} \end{aligned} \quad (1)$$

W_0 = Selected value of the initial value of initial weight

U_{ap} , U_{bp} , U_{cp} = in-phase unit templates.

In-phase unit templates are calculated by PCC phase voltage (V_{sa} , V_{sb} , V_{sc}). Relation between the phase voltage and amplitude of the PCC voltage (V_t). The amplitude of sensed PCC voltage is estimated as

$$V_t = \sqrt{2 \left(\frac{V_{sa}^2 + V_{sb}^2 + V_{sc}^2}{3} \right)} \quad (2)$$

The In-phase unit templates of PCC voltages (U_{ap} , U_{bp} , U_{cp}) are considered as,

$$\begin{aligned} U_{ap} &= V_{sa}/V_t; \\ U_{bp} &= V_{sb}/V_t \\ U_{cp} &= V_{sc}/V_t; \end{aligned} \quad (3)$$

The extracted values of I_{Lap} , I_{Lbp} , and I_{Lcp} is given to the sigmoid function as an activation function, and the outputs signals are (Z_{ap} , Z_{bp} , and Z_{cp}) of the feedforward section are expressed as

$$\begin{aligned} Z_{ap} &= f(I_{Lap}) = 1/(1 + e^{-I_{Lap}}) \\ Z_{bp} &= f(I_{Lbp}) = 1/(1 + e^{-I_{Lbp}}) \\ Z_{cp} &= f(I_{Lcp}) = 1/(1 + e^{-I_{Lcp}}) \end{aligned} \quad (4)$$

The estimated values of Z_{ap} , Z_{bp} , and Z_{cp} are fed to a hidden layer as input signals. The 3 phase outputs of this layer (I_{ap1} , I_{bp1} , and I_{cp1}) before the activation function are expressed as

$$I_{ap1} = W_{01} + W_{ap}Z_{ap} + W_{bp}Z_{bp} + W_{cp}Z_{cp}$$

International Journal for Research in Applied Science & Engineering Technology (IJRASET)

$$\begin{aligned} I_{bp1} &= W_{01} + W_{bp}Z_{bp} + W_{cp}Z_{cp} + W_{ap}Z_{ap} \\ I_{cp1} &= W_{01} + W_{cp}Z_{cp} + W_{ap}Z_{ap} + W_{bp}Z_{bp} \end{aligned} \quad (5)$$

Where W_{01} , W_{ap} , W_{bp} , and W_{cp} are the selected value of initial weight in hidden layer and the updated values of three phase weights using average weighted function value (W_p) of the active power current as a feedback signal.

The updated weights of phase “a” its active power components of load current “ W_{ap} ” at the n^{th} sampling instant is expressed as

$$W_{ap}(n) = W_p(n) + \mu [W_p(n) - W_{ap1}(n)] f'(I_{ap1}) Z_{ap}(n) \quad (6)$$

where $W_p(n)$ and $W_{ap}(n)$ are the average weighted value of the active power of load currents and the updated weighted value of phase “a” at the n^{th} sampling instant respectively. $W_{ap1}(n)$ and $Z_{ap}(n)$ are the phase “a” over all weighted value of the active power of the load current and the output of the feedforward of algorithm at the n^{th} instant, respectively. $f'(I_{ap1})$ and μ are represented as the derivative of I_{ap1} components and the learning rate.

Similarly, for phase “b” and phase “c,” the updated weighted values of the active power components of the load current are expressed as

$$\begin{aligned} W_{bp}(n) &= W_p(n) + \mu [W_p(n) - W_{bp1}(n)] f'(I_{bp1}) Z_{bp}(n) \\ W_{cp}(n) &= W_p(n) + \mu [W_p(n) - W_{cp1}(n)] f'(I_{cp1}) Z_{cp}(n) \end{aligned} \quad (7)$$

The extracted values of I_{ap1} , I_{bp1} , and I_{cp1} is given to the sigmoid function as an activation function to the estimation of the fundamental active components in terms of three phase weights W_{ap1} , W_{bp1} and W_{cp1}

$$\begin{aligned} W_{ap1} &= f(I_{ap1}) = 1/(1 + e^{-I_{ap1}}) \\ W_{bp1} &= f(I_{bp1}) = 1/(1 + e^{-I_{bp1}}) \\ W_{cp1} &= f(I_{cp1}) = 1/(1 + e^{-I_{cp1}}) \end{aligned} \quad (8)$$

The average weighted magnitude of the fundamental active power (W_p) is estimated using the amplitude sum of the 3 phase load active power (W_{ap1} , W_{bp1} and W_{cp1}) divided by three. It is required to generate load balancing, given as

$$W_p = (W_{ap1} + W_{bp1} + W_{cp1}) / 3 \quad (9)$$

First-order low-pass filters are used to separate the low frequency components. Where “k” denotes the scaled factor. This extracts the value of active power of current in this algorithm. After separation, low-frequency components are scaled to the actual value because the output of the activation function must lie between 0 and 1 and it is represented as W_{LpA} .

Similarly, the weighted amplitudes of the reactive power of the load currents (W_{aq} , W_{bq} , and W_{cq}) of the overall load current are extracted as

$$\begin{aligned} I_{Laq} &= W_0 + i_{La}U_{aq} + i_{Lb}U_{bq} + i_{Lc}U_{cq} \\ I_{Lbq} &= W_0 + i_{Lb}U_{bq} + i_{Lc}U_{cq} + i_{La}U_{aq} \\ I_{Lcq} &= W_0 + i_{Lc}U_{cq} + i_{La}U_{aq} + i_{Lb}U_{bq} \end{aligned} \quad (10)$$

Where W_0 is the selected values of starting weight and U_{aq} , U_{bq} , and U_{cq} are the quadrature components of the unit template. This unit templates (U_{aq} , U_{bq} and U_{cq}) of the phase Point of Common Coupling voltage are estimated as

$$\begin{aligned} U_{aq} &= (-3U_{ap} + U_{bp} + U_{cp})/\sqrt{3}; \\ U_{bq} &= (3U_{ap} + U_{bp} - U_{cp})/\sqrt{3}; \\ U_{cq} &= (-3U_{ap} + U_{bp} - U_{cp})/\sqrt{3}; \end{aligned} \quad (11)$$

The extracted values of I_{Laq} , I_{Lbq} , and I_{Lcq} is given to the sigmoid function as the activation function to the estimation of Z_{aq} , Z_{bq} , and Z_{cq}

$$\begin{aligned} Z_{aq} &= f(I_{Laq}) = 1/(1 + e^{-I_{Laq}}) \\ Z_{bq} &= f(I_{Lbq}) = 1/(1 + e^{-I_{Lbq}}) \end{aligned} \quad (12)$$

International Journal for Research in Applied Science & Engineering Technology (IJRASET)

$$Z_{cq} = f(I_{Lcq}) = 1/(1 + e^{-I_{Lcq}})$$

The estimated values of Z_{aq} , Z_{bq} , and Z_{cq} are given to the hidden layer as input signals. The 3 phase outputs of this layer is (I_{aq1} , I_{bq1} , and I_{cq1}) before the activation function can be represented as

$$\begin{aligned} I_{aq1} &= W_{01} + W_{aq}Z_{aq} + W_{bq}Z_{bq} + W_{cq}Z_{cq} \\ I_{bq1} &= W_{01} + W_{bq}Z_{bq} + W_{cq}Z_{cq} + W_{aq}Z_{aq} \\ I_{cq1} &= W_{01} + W_{cq}Z_{cq} + W_{aq}Z_{aq} + W_{bq}Z_{bq} \end{aligned} \quad (13)$$

Where W_{01} , W_{aq} , W_{bq} , and W_{cq} are the selected value that is starting weight in hidden layer and the updated three weights by the average weighted value of the reactive power (W_q) as a feedback signal, respectively

The updated weight of the single phase “a” reactive power of load currents “ W_{aq} ” at the n th sampling instant is expressed as

$$W_{aq}(n) = W_q(n) + \mu [W_q(n) - W_{aq1}(n)] f'(I_{aq1}) Z_{aq}(n) \quad (14)$$

$W_q(n)$ and $W_{aq}(n)$ are the average weighted value of reactive power of load currents and weight in the n th sampling instant and $W_{aq1}(n)$, $Z_{aq}(n)$ are the single phase “a” weighted amplitude of the reactive power current of load currents and the output of the feedforward of the algorithm at the n th instant, respectively. $f'(I_{aq1})$ and μ are represented as the derivative of I_{aq1} components and the learning rate.

Similarly, for phase “b” and “c,” the updated weighted values of the reactive power current of the load current is expressed as

$$\begin{aligned} W_{bq}(n) &= W_q(n) + \mu [W_q(n) - W_{bq1}(n)] f'(I_{bq1}) Z_{bq}(n) \\ W_{cq}(n) &= W_q(n) + \mu [W_q(n) - W_{cq1}(n)] f'(I_{cq1}) Z_{cq}(n) \end{aligned} \quad (15)$$

The extracted values of I_{aq1} , I_{bq1} , and I_{cq1} given to the activation function to the estimation of the fundamental reactive component in terms of three phase weights W_{aq1} , W_{bq1} , and W_{cq1} as

$$\begin{aligned} W_{aq1} &= f(I_{aq1}) = 1/(1 + e^{-I_{aq1}}) \\ W_{bq1} &= f(I_{bq1}) = 1/(1 + e^{-I_{bq1}}) \\ W_{cq1} &= f(I_{cq1}) = 1/(1 + e^{-I_{cq1}}) \end{aligned} \quad (16)$$

The average weight of the amplitudes of the fundamental reactive power components (W_q) is estimated using the amplitude summation of the three phase load reactive power components of the load current (W_{aq1} , W_{bq1} , and W_{cq1}) divided by three. Mathematically, it is expressed as

$$W_q = (W_{aq1} + W_{bq1} + W_{cq1}) / 3 \quad (17)$$

First-order low-pass filters are used to separate the low frequency component. “ r ” denotes the scaled factor. This will have the extracted value of reactive power components in the algorithm. After separating low-frequency components and scaling to the actual value because the output of the activation function is between 0 and 1, it is represented as $WLqA$.

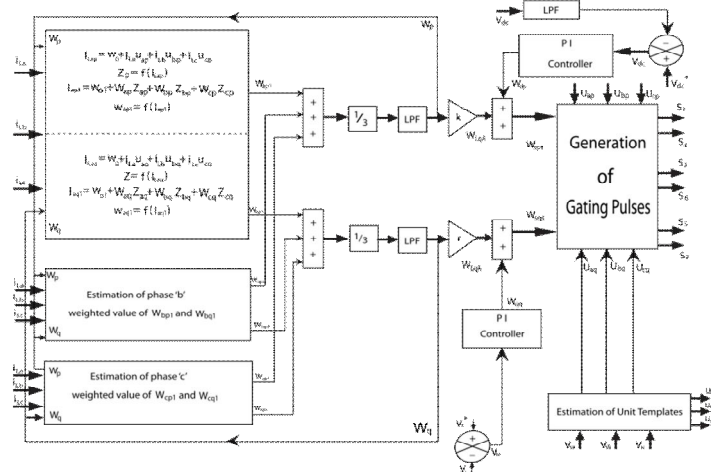


Fig 3. Estimating Reference Current using BP Control Algorithm

International Journal for Research in Applied Science & Engineering Technology (IJRASET)

B. Active Power Current Components Of Reference Source Current

An error in the dc voltage is obtained by comparing the reference dc bus voltage V_{dc}^* and the sensed dc bus voltage V_{dc} of a VSC, and this error at the n^{th} sampling instant is expressed as

$$V_{de}(n) = V_{dc}^*(n) - V_{dc}(n) \quad (18)$$

This error voltage is fed to a proportional Integral (PI) controller whose output is required for maintaining the dc bus voltage of the DSTATCOM. At the n^{th} sampling instant, the output of the PI controller is as follows:

$$W_{dp}(n) = W_{dp}(n-1) + K_{pd}[V_{de}(n) - V_{de}(n-1)] + K_{id}V_{de}(n) \quad (19)$$

Where k_{pd} and k_{id} are the proportional and integral gain constants of the dc bus PI controller. $V_{de}(n)$ and $V_{de}(n-1)$ are the dc bus voltage errors in the n^{th} and $(n-1)^{\text{th}}$ instant, and $W_{dp}(n)$ and $W_{dp}(n-1)$ are the amplitudes of the active power component of fundamental reference current at the n^{th} and $(n-1)^{\text{th}}$ instant. The amplitude of the active power current components of the reference source current (W_{spt}) is estimated by the addition of the output of the dc bus PI controller (W_{dp}) and the average magnitude of the load active currents (W_{LpA}) as

$$W_{spt} = W_{dp} + W_{LpA} \quad (20)$$

C. Reactive Power Component Of Reference Source Currents

An error in the ac bus voltage is achieved after comparing the amplitudes of the reference ac bus voltage v_i^* and the sensed ac bus voltage v_i of a VSC. The extracted ac bus voltage error v_{ie} at the n^{th} sampling instant is expressed as

$$V_{ie}(n) = V_i^*(n) - V_i(n) \quad (21)$$

The weighted output of the ac bus PI controller W_{qq} for regulating the ac bus terminal voltage at the n^{th} sampling instant is expressed as

$$W_{qq}(n) = W_{qq}(n-1) + K_{pi}[V_{ie}(n) - V_{ie}(n-1)] + K_{it}V_{ie}(n) \quad (22)$$

Where $W_{qq}(n)$ is part of the reactive power component of the source current and it is renamed as W_{qq} . K_{pi} and k_{it} are the proportional Integral constants gain of the ac bus voltage PI controller. The amplitude of the reactive power current components of the reference source current (W_{sqt}) is calculated by subtracting the output of the voltage PI controller (W_{qq}) and the average load reactive currents (W_{LqA}) as

$$W_{sqt} = W_{qq} - W_{LqA} \quad (23)$$

D. Estimating Reference Source Current And Generating Pulses For IGBT

Three phase reference source current that is active and reactive current components are estimated using the amplitude of three phase (a, b, and c) load active current components, Point of Common Coupling voltage In-phase unit templates, reactive power current components and Point of Common Coupling voltage unit templates as

$$\begin{aligned} i_{sap} &= W_{spt}U_{ap}; i_{sbp} = W_{spt}U_{bp}; i_{scp} = W_{spt}U_{cp} \\ i_{saq} &= W_{sqt}U_{aq}; i_{sbq} = W_{sqt}U_{bq}; i_{scq} = W_{sqt}U_{cq} \end{aligned} \quad (24)$$

The summation result of the reference active and reactive current components gives the reference source currents and these are given as

$$\begin{aligned} i_{sa}^* &= i_{sap}^* + i_{saq}^* \\ i_{sb}^* &= i_{sbp}^* + i_{sbq}^* \\ i_{sc}^* &= i_{scp}^* + i_{scq}^* \end{aligned} \quad (25)$$

The sensed source currents (i_{sa} , i_{sb} , i_{sc}) and the reference source currents (i_{sa}^* , i_{sb}^* , i_{sc}^*) are compared, and error signals are being amplified through an PI current regulators. The output from the controller is fed to pulse width modulation (PWM) controller to generate the gating signals for insulated-gate bipolar transistors (IGBTs) S1 to S6 of the VSC used as a DSTATCOM.

IV. SIMULATION RESULTS AND DISCUSSION

The performance of the VSC-DSTATCOM by using the BP control algorithm is given below. A three phase source is connected in series with linear and nonlinear load through the transmission line elements such as two three phase transformer. One of the transformer is star-star connected whereas another will be delta to star grounded. Distributed Static Compensator is connected at the middle of the transmission line for Reactive Power Compensation. It can either absorb or generates reactive power and injected into

International Journal for Research in Applied Science & Engineering Technology (IJRASET)

the transmission line with the help of coupling transformer. DSTATCOM consists of six power semiconductor devices for which gate pulses are generated using Back Propagation Control Algorithm.

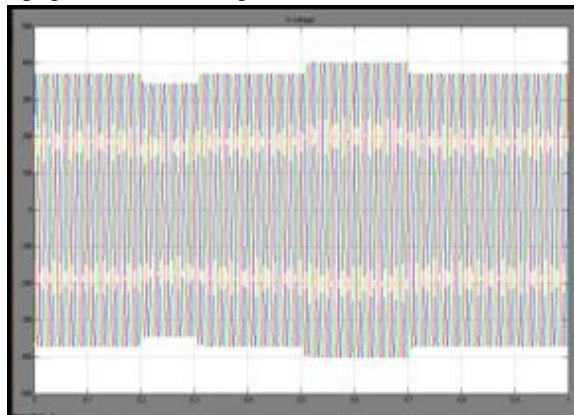


Fig 4. Voltage swell and Sag versus Time without DSTATCOM and Control Algorithm

The Output waveform shown in Fig 4 of voltage sag in three phase and single phase (V_a, V_b, V_c) represented as voltage versus time characteristics. Due to disconnection of large load through the circuit breaker, there will be sag in the voltage on the distribution system and the output waveform of voltage swell in three phase and single phase (V_a, V_b, V_c) represented as voltage versus time characteristics. Due to connection of large load through the circuit breaker, there will be voltage swell on the distribution system. From 0.2 to 0.3 seconds there will be voltage sag and 0.5 to 0.7 there will voltage swell due to connection and disconnection of the loads which is shown in Fig 4.

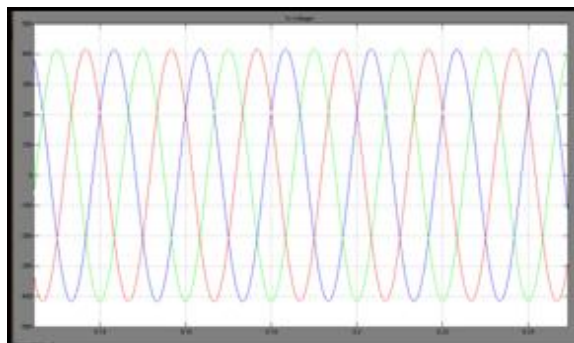


Fig 5. Three Phase Output Voltage versus Time with DSTATCOM and Control Algorithm

The output waveform shown in Fig 5 of compensated voltage using DSTATCOM in three phase and single phase (V_a, V_b, V_c) represented as voltage versus time characteristics. DSTATCOM either can absorb or generate reactive power in the transmission line in order to obtain constant voltage profile throughout the system.

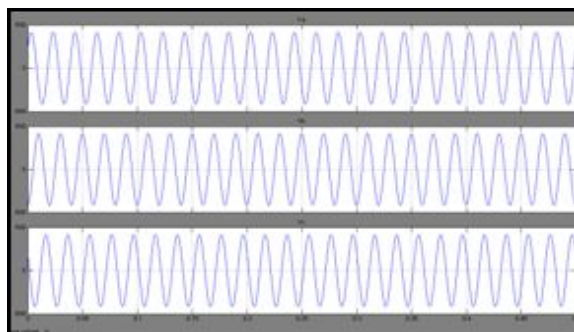


Fig 6. Three phase output voltage with versus Time DSTATCOM and Control Algorithm

International Journal for Research in Applied Science & Engineering Technology (IJRASET)

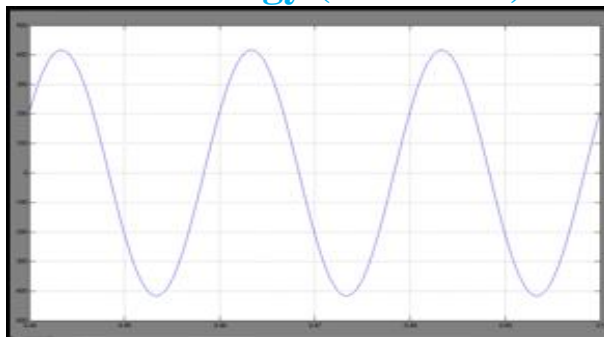


Fig 7. Output voltage verses Time for one phase with DSTATCOM and Control Algorithm

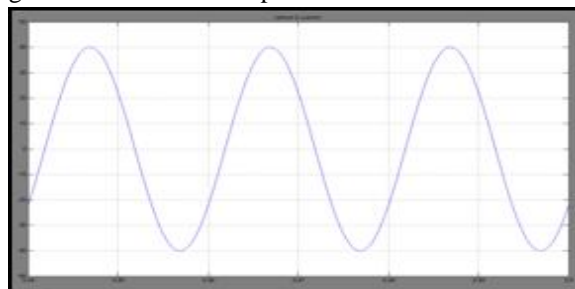


Fig 8. Output current verses Time for one phase Current

Output current for one phase is 40 A which is shown in Figure 8.

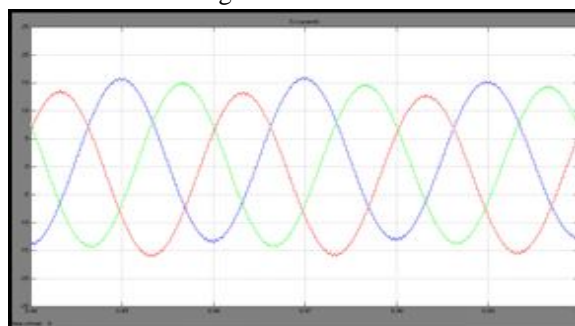


Fig 9. Waveform for compensation current verses Time

Tabular column

Table 1

This table 1formation shows the comparison between voltage swell and voltage sag with and without BP control Algorithm. Table 2 represents the system specifications which are given below.

Conditions	Source voltage	Load voltage	Voltage swell	Voltage sag
Without DSTATCOM	415	380	400 (swell)	340 (sag)
With DSTATCOM	415	415	Absence of swell	Absence of swell

International Journal for Research in Applied Science & Engineering Technology (IJRASET)

V. CONCLUSION

A most preferred solution for the power quality improvement using a VSC-DSTATCOM. This paper has proposed the development of DSTATCOM and BP training algorithm for power quality improvement in a three phase three wire distribution system. BP training algorithm has been used for the extraction of reference source currents to generate the pulses for IGBT of the VSC-DSTATCOM. Back Propagation (BP) Algorithm improves the speed of convergence and reduces the oscillations during weight updating action. Most preferred solution for Power quality improvement as to maintain rated PCC voltage. From the result it has been found that, this DSTATCOM and its control algorithm is well satisfied for reactive power compensation. VSC-DSTATCOM has quick response time and accurate compensation effect.

REFERENCES

- [1] Jos Arrillaga and Neville R. Watson, "power system harmonics", John Wiley Sons Ltd, (2003).
- [2] A. Ortiz, C. Gherasim, M. Manana, C. J. Renedo, L. I. Eguiluz, and R. J. M. Belmans, "Total harmonic distortion decomposition depending on distortion origin," IEEE Trans. Power Del., Vol. 20, No. 4, pp. 2651–2656, Oct. 2005.
- [3] IEEE Recommended Practices and Requirement for Harmonic Control on "Electric Power System," IEEE Std. 519, 1992.
- [4] T.-L. Lee, S.-H. Hu, and Y.-H. Chan, "DSTATCOM with positive sequence admittance and negative sequence conductance to mitigate voltage fluctuations in high-level penetration of distributed generation systems," IEEE Trans. Ind. Electron., Vol. 60, No. 4, pp. 1417–1428, Apr. 2013.
- [5] G. Benysek and M. Pasko, "Power Theories for Improved Power Quality," London, U.K.: Springer-Verlag, 2012.
- [6] B. Singh and J. Solanki, "A comparison of control algorithms for DSTATCOM," IEEE Trans. Ind. Electron., Vol. 56, No. 7, pp. 2738–2745, Jul. 2009.
- [7] C. H. da Silva, R. R. Pereira, L. E. B. da Silva, G. Lambert-Torres, B. K. Bose, and S. U. Ahn, "A digital PLL scheme for three-phase system using modified synchronous reference frame," IEEE Trans. Ind. Electron., Vol. 57, No. 11, pp. 3814–3821, Nov. 2010.
- [8] Bhmsingh, Jitendra Solanki "A Comparison of Control Algorithms for DSTATCOM", IEEE Transactions On Industrial Electronics, Vol. 56, No. 7, July 2009.
- [9] L. L. Lai, W. L. Chan, and A. T. P. So, "A two-ANN approach to frequency and harmonic evaluation," in Proc. 5th Int. Conf. Artif. Neural Netw., 1997.
- [10] X. Mao, "The harmonic currents detecting algorithm based on adaptive neural network," in Proc. 3rd Int. Symp. Intell. Inf. Technol. Appl., 2009, Vol. 3.
- [11] A. Zouidi, F. Fnaiech, K. Al-Haddad, and S. Rahmani, "Artificial neural networks as harmonic detectors," in Proc. 32nd Annu. Conf. IEEE Ind. Electron., 2006, pp. 2889–2892.



10.22214/IJRASET



45.98



IMPACT FACTOR:
7.129



IMPACT FACTOR:
7.429



INTERNATIONAL JOURNAL FOR RESEARCH

IN APPLIED SCIENCE & ENGINEERING TECHNOLOGY

Call : 08813907089  (24*7 Support on Whatsapp)